

# ME 433 – STATE SPACE CONTROL

## Lecture 4

## State Transformation

We consider the linear, time-invariant system

$$\begin{aligned}\dot{x} &= Ax + Bu, \\ y &= Cx + Du.\end{aligned}$$

We define the state transformation

$$x(t) = Tz(t) \Leftrightarrow T^{-1}x(t) = z(t)$$

Then we can write

$$\begin{aligned}T\dot{z} &= ATz + Bu \Rightarrow \dot{z} = T^{-1}ATz + T^{-1}Bu \\ y &= CTz + Du.\end{aligned}$$

to obtain

$$\begin{aligned}\dot{z} &= \tilde{A}z + \tilde{B}u \\ y &= \tilde{C}z + \tilde{D}u\end{aligned} \quad \boxed{\tilde{A} = T^{-1}AT, \tilde{B} = T^{-1}B, \tilde{C} = CT, \tilde{D} = D}$$

The state-space representation is NOT unique!

## Solution of State Equation

### Time-invariant Dynamics:

We consider the linear, time-invariant, homogeneous system

$$\dot{x} = Ax + Bu$$

where  $A$  is a constant  $n \times n$  matrix. The solution can be written as

$$x(t) = e^{A(t-\tau)}x(\tau) + \int_{\tau}^t e^{A(t-\lambda)}Bu(\lambda)d\lambda$$

### Time-variant Dynamics:

We consider the linear, time-variant, homogeneous system

$$\dot{x} = A(t)x + Bu$$

where  $A$  is a time-variant  $n \times n$  matrix. The solution can be written as

$$x(t) = \Phi(t, \tau)x(\tau) + \int_{\tau}^t \Phi(t, \lambda)B(\lambda)u(\lambda)d\lambda$$

## Transfer Function

We consider the linear, time-invariant system

$$\dot{x} = Ax + Bu,$$

$$y = Cx + Du.$$

We Laplace transform the state equation to obtain

$$sX(s) - x(0) = AX(s) + BU(s)$$

$$L[\dot{x}(t)] = sX(s) - x(0)$$

$$X(s) = L[x(t)] \quad U(s) = L[u(t)]$$

And we solve to obtain

$$X(s) = (sI - A)^{-1}x(0) + (sI - A)^{-1}BU(s)$$

We Laplace transform the output equation to obtain

$$Y(s) = CX(s) + DU(s)$$

## Solution of State Equation

$$X(s) = (sI - A)^{-1} x(0) + (sI - A)^{-1} B U(s)$$

We inverse Laplace transform to obtain

$$x(t) = e^{At} x_0 + \int_0^t e^{A(t-\tau)} B u(\tau) d\tau$$

where  $L[e^{At}] = (sI - A)^{-1}$ ;  $L\left[\int_0^t f(t-\tau)g(\tau)d\tau\right] = f(s)g(s)$

For LTI systems, the exponential matrix  $e^{At}$  is the transition matrix of the system. Its Laplace transform  $\Phi(s) = (sI - A)^{-1}$  is called the resolvent of the system. Faddeeva's algorithm allows for a fast computation of  $\Phi(s)$ .

How can we compute the transition matrix  $e^{At}$ ?

1. Compute  $sI - A$ .
2. Obtain the resolvent  $\Phi(s)$  by inverting  $sI - A$ .
3. Obtain the transition matrix by computing the inverse Laplace transform of the resolvent, element by element.

## Transfer Function

Assuming  $x(0)=0$  and combining state and output equations we obtain the transfer function

$$Y(s) = [C(sI - A)^{-1} B + D]U(s)$$

The transfer function does NOT depend on the state choice because it represents the input-output relationship.

Given two state space representations

$$\begin{array}{ll} \dot{x} = Ax + Bu, & \dot{z} = \tilde{A}z + \tilde{B}u \\ y = Cx + Du. & y = \tilde{C}z + \tilde{D}u \end{array}$$

$$Y(s) = [\tilde{C}(sI - \tilde{A})^{-1} \tilde{B} + \tilde{D}]U(s) = [C(sI - A)^{-1} B + D]U(s)$$

## Controllability and Observability

Let us consider the system

$$\dot{x} = Ax + Bu,$$

$$y = Cx + Du.$$

with

$$A = \begin{bmatrix} 2 & 3 & 2 & 1 \\ -2 & -3 & 0 & 0 \\ -2 & -2 & -4 & 0 \\ -2 & -2 & -2 & -5 \end{bmatrix}, B = \begin{bmatrix} 1 \\ -2 \\ 2 \\ -1 \end{bmatrix}, C = \begin{bmatrix} 7 & 6 & 4 & 2 \end{bmatrix}, D = 0$$

and transfer function

$$H(s) = C(sI - A)^{-1}B + D = \frac{s^3 + 9s^2 + 26s + 24}{s^4 + 10s^3 + 35s^2 + 50s + 4}$$

## Controllability and Observability

The transfer function can be written as

$$H(s) = \frac{(s+2)(s+3)(s+4)}{(s+1)(s+2)(s+3)(s+4)} = \frac{1}{(s+1)}$$

What's going on? Let's see the system from a different angle. Consider:

$$x(t) = Tz(t) \Leftrightarrow T^{-1}x(t) = z(t)$$

with

$$T = \begin{bmatrix} 1 & -1 & 0 & 0 \\ -1 & 2 & -1 & 0 \\ 0 & -1 & 2 & -1 \\ 0 & 0 & -1 & 2 \end{bmatrix}$$

to obtain

## Controllability and Observability

$$\dot{z} = \tilde{A}z + \tilde{B}u$$

$$y = \tilde{C}z + \tilde{D}u$$

with

$$\tilde{A} = T^{-1}AT = \begin{bmatrix} -1 & 0 & 0 & 0 \\ 0 & -2 & 0 & 0 \\ 0 & 0 & -3 & 0 \\ 0 & 0 & 0 & -4 \end{bmatrix},$$

$$\tilde{B} = T^{-1}B = \begin{bmatrix} 1 \\ 0 \\ 1 \\ 0 \end{bmatrix}, \tilde{C} = CT = [1 \ 1 \ 0 \ 0] \tilde{D} = 0$$

## Controllability and Observability

Let us assume the following modal form:

$$\dot{x} = Ax + Bu,$$

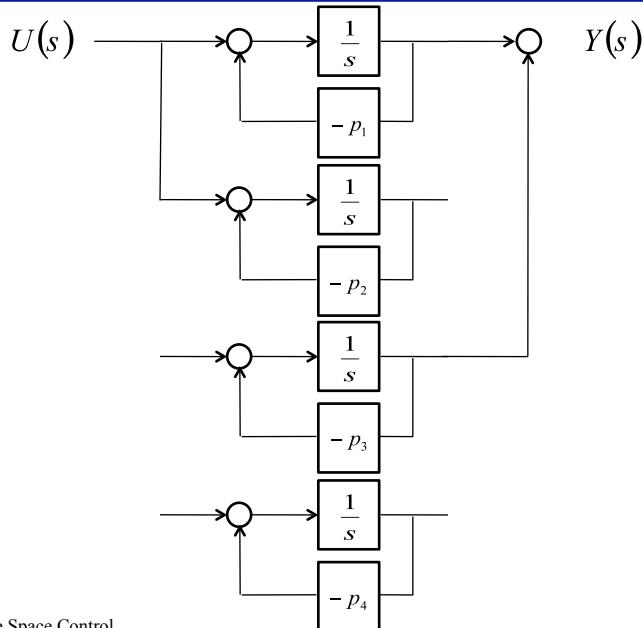
$$y = Cx + Du.$$

$$A = \begin{bmatrix} -p_1 & 0 & 0 & 0 \\ 0 & -p_2 & 0 & 0 \\ 0 & 0 & -p_3 & 0 \\ 0 & 0 & 0 & -p_4 \end{bmatrix}, B = \begin{bmatrix} 1 \\ 1 \\ 0 \\ 0 \end{bmatrix}, C = [k_1 \ 0 \ k_3 \ 0] D = 0$$

$$Y(s) = \frac{k_1}{s + p_1} U(s) + \frac{0}{s + p_2} U(s) + \frac{k_3}{s + p_3} 0 + \frac{0}{s + p_4} 0$$

Only the controllable and observable mode appears in the transfer function

## Controllability and Observability



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## Controllability

**Problem Definition:** "A system is said to be controllable if and only if it is possible, by means of the input, to transfer the system from any initial state  $x(0)$  to any other state  $x(t)$  in a finite time  $t \geq 0$ "

$$x(t) = e^{At}x(0) + \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau$$

Let us take

$$u(\tau) = B^T e^{A^T(t-\tau)} P^{-1}(t) [x(t) - e^{At}x(0)]$$

where

$$P(t) = \int_0^t e^{A(t-\tau)} B B^T e^{A^T(t-\tau)} d\tau$$

is invertible.

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## Controllability

Replacing  $u$  in the expression for  $x$  we will demonstrate that this is an input that drives the state from  $x(0)$  to  $x(t)$ .

$$\begin{aligned} x(t) &= e^{At}x(0) + \int_0^t e^{A(t-\tau)}B\left[B^T e^{A^T(t-\tau)}P^{-1}(t)[x(t) - e^{At}x(0)]\right]d\tau \\ &= e^{At}x(0) + \int_0^t e^{A(t-\tau)}BB^T e^{A^T(t-\tau)}d\tau P^{-1}(t)[x(t) - e^{At}x(0)] \\ &= e^{At}x(0) + P(t)P^{-1}(t)[x(t) - e^{At}x(0)] \\ &= x(t) \end{aligned}$$

It is indeed possible to show that this is the only input that drives the state from  $x(0)$  to  $x(t)$ .

**Theorem:** A system is controllable if and only if the matrix

$$P(t) = \int_0^t e^{A(t-\tau)}BB^T e^{A^T(t-\tau)}d\tau \quad \text{Controllability Gramian}$$

is nonsingular for some  $t \geq 0$ .

**Proof:** In class

## Controllability

The time derivative of the Controllability Gramian

$$P(t) = \int_0^t e^{A(t-\tau)}BB^T e^{A^T(t-\tau)}d\tau$$

is given by (Leibniz's rule)

$$\begin{aligned} \frac{d}{dt}P(t) &= e^{A(t-\tau)}BB^T e^{A^T(t-\tau)} \Big|_{\tau=t} + \\ &+ \int_0^t \frac{d}{dt} \left[ e^{A(t-\tau)} \right] BB^T e^{A^T(t-\tau)} d\tau + \int_0^t e^{A(t-\tau)}BB^T \frac{d}{dt} \left[ e^{A^T(t-\tau)} \right] d\tau \\ &= BB^T + \int_0^t \frac{d}{dt} \left[ e^{A(t-\tau)} \right] BB^T e^{A^T(t-\tau)} d\tau + \int_0^t e^{A(t-\tau)}BB^T \frac{d}{dt} \left[ e^{A^T(t-\tau)} \right] d\tau \end{aligned}$$

Then, the Controllability Gramian satisfies

$$\dot{P}(t) = BB^T + AP(t) + P(t)A^T, \quad P(0) = 0 \quad \text{Lyapunov Equation}$$

## Controllability

**Theorem:** “A system is controllable if and only if the matrix

$$\bar{C} = \begin{bmatrix} B & AB & A^2B & \cdots & A^{n-1}B \end{bmatrix} \text{ Controllability Matrix}$$

is full-rank.”

Proof: In class

## Observability

**Problem Definition:** “An unforced system is said to be observable if and only if it is possible to determine any (arbitrary initial) state  $x(0)$  by using only a finite record,  $y(\tau)$  for  $0 \leq \tau \leq T$ , of the output”

**Theorem:** “A system is observable if and only if the matrix

$$Q(t) = \int_0^t e^{A^T(t-\tau)} C^T C e^{A(t-\tau)} d\tau \quad \text{Observability Gramian}$$

is nonsingular for some  $t \geq 0$ .”

The Observability Gramian satisfies

$$\dot{Q}(t) = C^T C + A^T Q(t) + Q(t) A, \quad Q(0) = 0 \quad \text{Lyapunov Equation}$$

**Theorem:** “A system is observable if and only if the matrix

$$\bar{O} = \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{bmatrix} \quad \text{Observability Matrix}$$

is full-rank.”

## Controllability & Observability

### **Theorem (Popov-Belevitch-Hautus): Eigenvector Tests**

A pair  $\{A,B\}$  will be noncontrollable if and only if there exists a row vector  $q \neq 0$  such that

$$qA = \lambda q, \quad qB = 0$$

In other words,  $\{A,B\}$  will be controllable if and only if there is no row (or left) eigenvector of  $A$  that is orthogonal to  $B$ .

A pair  $\{C,A\}$  will be nonobservable if and only if there exists a column vector  $p \neq 0$  such that

$$Ap = \lambda p, \quad Cp = 0$$

In other words,  $\{C,A\}$  will be observable if and only if there is no column (or right) eigenvector of  $A$  that is orthogonal to  $C$ .

Proof: In class

## Controllability & Observability

### **Theorem (Popov-Belevitch-Hautus): Rank Tests**

A pair  $\{A,B\}$  will be controllable if and only if

$$\text{rank} \begin{bmatrix} sI - A & B \end{bmatrix} = n \quad \text{for all } s$$

A pair  $\{C,A\}$  will be observable if and only if

$$\text{rank} \begin{bmatrix} C \\ sI - A \end{bmatrix} = n \quad \text{for all } s$$

Proof: In class

## **Controllability & Observability**

Examples :