

ME 343 – Control Systems

Lecture 23

October 16, 2009

Frequency Response

$$u(t) = A \cos(\omega t + \phi) \rightarrow \boxed{G(s)} \rightarrow y_{ss} = |G(j\omega)| A \cos(\omega t + \phi + \angle G(j\omega))$$

Stable Transfer Function

- After a transient, the output settles to a sinusoid with an amplitude magnified by $|G(j\omega)|$ and phase shifted by $\angle G(j\omega)$.
- Since all signals can be represented by sinusoids (Fourier series and transform), the quantities $|G(j\omega)|$ and $\angle G(j\omega)$ are extremely important.
- Bode developed methods for quickly finding $|G(j\omega)|$ and $\angle G(j\omega)$ for a given $G(s)$ and for using them in control design.

Bode Diagrams

$$G(s) = K \frac{(s - z_1)(s - z_2) \cdots (s - z_m)}{(s - p_1)(s - p_2) \cdots (s - p_n)}$$



$$G(s) = \left| K \right| \frac{r_1^z r_2^z \cdots r_m^z}{r_1^p r_2^p \cdots r_n^p} e^{i[\phi^K + (\phi_1^z + \phi_2^z + \cdots + \phi_m^z) - (\phi_1^p + \phi_2^p + \cdots + \phi_n^p)]}$$

The magnitude and phase of $G(s)$ when $s=j\omega$ is given by:

Nonlinear in the magnitudes

$$|G(j\omega)| = \left| K \right| \frac{r_1^z r_2^z \cdots r_m^z}{r_1^p r_2^p \cdots r_n^p},$$

$$\angle G(j\omega) = \phi^K + (\phi_1^z + \phi_2^z + \cdots + \phi_m^z) - (\phi_1^p + \phi_2^p + \cdots + \phi_n^p)$$

Linear in the phases

Bode Diagrams

Why do we express $|G(j\omega)|$ in decibels?

$$|G(j\omega)|_{dB} = 20 \log |G(j\omega)|$$

$$|G(j\omega)| = \left| K \right| \frac{r_1^z r_2^z \cdots r_m^z}{r_1^p r_2^p \cdots r_n^p} \Rightarrow |G(j\omega)|_{dB} = ?$$

By properties of the logarithm we can write:

$$20 \log |G(s)| = 20 \log |K| + (20 \log r_1^z + 20 \log r_2^z + \cdots + 20 \log r_m^z) - (20 \log r_1^p + 20 \log r_2^p + \cdots + 20 \log r_n^p)$$

The magnitude and phase of $G(s)$ when $s=j\omega$ is given by:

Linear in the magnitudes (dB)

$$|G(s)|_{dB} = \left| K \right|_{dB} + (r_1^z|_{dB} + r_2^z|_{dB} + \cdots + r_m^z|_{dB}) - (r_1^p|_{dB} + r_2^p|_{dB} + \cdots + r_n^p|_{dB})$$

$$\angle G(s) = \phi^K + (\phi_1^z + \phi_2^z + \cdots + \phi_m^z) - (\phi_1^p + \phi_2^p + \cdots + \phi_n^p)$$

Linear in the phases

General Transfer Function (Bode Diagrams)

$$G(j\omega) = K_o (j\omega)^m (j\omega\tau + 1)^n \left[\left(\frac{j\omega}{\omega_n} \right)^2 + 2\zeta \frac{j\omega}{\omega_n} + 1 \right]^q$$

The **magnitude (dB)** (**phase**) is the sum of the **magnitudes (dB)** (**phases**) of each one of the terms. We learn how to plot each term, we learn how to plot the whole magnitude and phase Bode Plot.

Classes of terms:

1- $G(j\omega) = K_o$

2- $G(j\omega) = (j\omega)^m$

3- $G(j\omega) = (j\omega\tau + 1)^n$

4- $G(j\omega) = \left[\left(\frac{j\omega}{\omega_n} \right)^2 + 2\zeta \frac{j\omega}{\omega_n} + 1 \right]^q$

Bode Diagrams

Example:

$$G(s) = \frac{2000(s + 0.5)}{s(s + 10)(s + 50)}$$

Frequency Response: Poles/Zeros in the RHP

- Same $|G(j\omega)|$.
- The effect on $\angle G(j\omega)$ is opposite than the stable case.

An unstable pole behaves like a stable zero
An "unstable" zero behaves like a "stable" pole

Example:

$$G(s) = \frac{1}{s-2}$$

This frequency response cannot be found experimentally
but can be computed and used for control design.